2. Transient and Frequency Analysis

- step response
- impulse response
- basic frequency analysis
- improved frequency analysis

Step response of a first-order system

$$G(s) = \frac{K}{1+sT}e^{-s\tau}, \quad y(t) = K\left[1 - e^{-(t-\tau)/T}\right]u(t-\tau)$$

determine K from the final value, T and τ from the steepest tangent



Step response of a second-order system

$$G(s) = \frac{K\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}$$
$$y(t) = K \left[1 - \frac{e^{-\zeta\omega_0 t}}{\sqrt{1 - \zeta^2}} \sin(\omega_0 \sqrt{1 - \zeta^2} t + \tau) \right], \quad \tau = \arccos \zeta$$



The local extrema of the step response occur at

$$t_{k} = \frac{k\pi}{\omega_{0}\sqrt{1-\zeta^{2}}}, \quad k = 1, 2, \dots \text{ and } y(t_{k}) = K(1-(-1)^{k}M^{k})$$

$$K_{(1+M)}$$

$$K_{(1-M)}$$

$$t_{1}$$

$$t_{2}$$

$$t_{3}$$

$$t_{4}$$

determine ζ from

$$M = \exp\left(\frac{-\zeta\pi}{\sqrt{1-\zeta^2}}\right)$$

compute ω_0 when the period T of oscillations are determined

$$\omega_0 = \frac{2\pi}{T\sqrt{1-\zeta^2}}$$

Impulse response

consider a system described by

$$y(t) = \sum_{k=0}^{\infty} g(k)u(t-k) + v(t)$$

where v(t) is a disturbance and u(t) is an impulse input:

$$u(t) = \begin{cases} \alpha, & t = 0\\ 0, & t \neq 0 \end{cases}$$

if the noise level is low, the estimate of g(t) is

$$\hat{g}(t) = \frac{y(t)}{\alpha}$$

Basic frequency analysis



Sine-wave input: $u(t) = a \sin(\omega t)$

$$y(t) = a|G(i\omega)|\sin(\omega t + \phi) + \text{ transient}$$

- determine the amplitudes and the phase shift of y(t)
- repeat for a number of ω and obtain a graphical approximation of $G(\mathrm{i}\omega)$
- $\bullet\,$ in the presence of noise, it is difficult to estimate a and $\phi\,$

Determining amplitude and phase



Effect of noise



• example with a=2, G(s)=1/(s+1) at frequency $\omega_0=\sqrt{3}$

• noisy data make it difficult to determine amplitude and phase

Improved frequency analysis



suppress the effect of the noise e(t) by taking correlation with a cosine function

If $T=2k\pi/\omega$, then

$$y_s(T) = \frac{a|G(i\omega)|T}{2}\cos\phi + \int_0^T e(t)\sin\omega t dt,$$
$$y_c(T) = \frac{a|G(i\omega)|T}{2}\sin\phi + \int_0^T e(t)\cos\omega t dt$$

- the integral terms can be considered as projections of e(t) on an orthonormal basis
- the estimate of $G(\mathrm{i}\omega)$ is

$$\operatorname{Re}\{\hat{G}(\mathrm{i}\omega)\} = \frac{2y_s(T)}{aT}, \quad \operatorname{Im}\{\hat{G}(\mathrm{i}\omega)\} = \frac{2y_c(T)}{aT}$$

References

Chapter 6 in L. Ljung, *System Identification: Theory for the User*, Prentice Hall, Second edition, 1999

Chapter 3 in

T. Söderström and P. Stoica, System Identification, Prentice Hall, 1989